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Distributed Control of Parallel DC-DC Converters Under FDI Attacks on Actuators

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Abstract—The parallel connection of DC-DC converters requires the development of an appropriate control strategy that regulates load voltage and shares current amongst participating converters. This paper proposes a resilient and robust cooperative distributed control approach that simultaneously ensures voltage regulation and balanced current sharing in parallel DC-DC converters in the presence of false data injection attacks on control input channels. Based on analytical tools from network control and Lyapunov stability theory, concise stability certificates are derived. The proposed cooperative distributed control strategy guarantees resilience against unknown bounded attacks on the actuators of DC-DC converters and the robustness to uncertainties in load parameters and the physical parameters of converters. Furthermore, the control design for each converter does not require any knowledge about the number of participating converters. The detailed simulation and experimental results verify the satisfactory performance of the proposed method in voltage regulation and balanced current sharing in parallel converters, as well as resilience to bounded false data injection attacks.

Index Terms—Parallel DC-DC converters, cooperative distributed control, resilient control, false data injection (FDI) attacks.

I. INTRODUCTION

THE parallel-connected DC-DC converter systems offer several advantages over a single high-capacity centralized converter, including increased reliability, potential for higher efficiency, better dynamic performance, ease of maintenance and repair, improved thermal management, and reduced stress levels on the constituent converters, as the total load current can be shared among the converters [1]–[3]. Due to their numerous advantages, they have been extensively used in a large number of applications such as railway vehicles, electric aircraft, and zero-emission ferries.

Despite the potential benefits that the parallel interconnection of DC-DC converters bring, they require appropriate control schemes to regulate load voltage and accurately share load demands amongst existing converters [2]. Unbalanced current

distribution causes the converter overloading and overheating and might lead to the overall system’s failure [4]. Extensive research has been carried out in the area of designing controls for accurate load sharing in parallel DC-DC converters, e.g. droop-based methods [5]–[8], integral-variable-structure- and multiple-sliding-surface-based control [9], master-slave current sharing control [10], finite-time control [11], geometric decoupling in state and input spaces [12], [13], as well as cooperative and distributed control techniques [4], [14]–[16]. These approaches assume an ideal control framework with ideal sensors, actuators, and communication networks. Nevertheless, such an assumption might not be realistic, as in practice communication links can fail and cyber-attacks can easily compromise the normal operation of control systems. These events might lead to detrimental impacts on the stability and performance of entire systems. Since parallel DC-DC converters are often used in mission-critical applications where cybersecurity and reliability are a main concern, it is essential to enhance the resilience of control systems against cyber-attacks and infiltration.

False data injection (FDI) attacks are one of the most common cyber-attacks which compromise control systems by injecting false information into their vulnerable elements; i.e. sensors, actuators, or communication links [17]. To enhance resilience in DC systems against FDI attacks, several model-based and data-driven attack detection and mitigation methods have recently been investigated in [18]–[22] and references therein. However, these approaches mainly rely on a strict assumption that at least half of the disrupted converters’ neighbors should be healthy. As a result, they are not applicable for worst-case scenarios in which all DC-DC converters are subject to FDI cyber-attacks. To deal with this limitation, resilient distributed control strategies have emerged. A trust-based cooperative control strategy for DC microgrids under FDI cyber-attacks on communication links and controller hijacking has been proposed in [23]. Although this control paradigm mitigates the adverse effects of such attacks, it entails high computation burdens in order to calculate the trustworthiness of incoming information at each converter. Moreover, to ensure an attack-resilient operation of DC networks using the proposed control approach in [23], more than half of the neighboring converters should be healthy and not be under cyber-attacks. A resilient distributed adaptive control mechanism against unbounded FDI attacks on actuators has been developed in [24]. Since in practice bounded attacks are more probable [25], as any unbounded attacks can easily be

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detected by an anomaly detection protocol, the relevance of the problem presented in [24] is somewhat limited.

To the best of our knowledge, the research on cyber-attack-resilience of control mechanisms in parallel DC-DC converters is still in its infancy and can benefit from further studies. Motivated by this and the aforementioned concerns with the existing attack detection techniques, this paper presents a cooperative distributed control framework developed for parallel DC-DC converters, paying special attention to the resilience of distributed controllers against FDI cyber-attacks on control input channels (actuators). This type of cyber-attacks makes control decisions incorrect, potentially causing equipment damage and adversely impacting the system performance. The main contributions of this paper can be summarized as follows:

- In contrast to conventional cooperative and distributed control strategies (e.g., [4], [11], [14]–[16]), the proposed distributed control framework is resilient to FDI cyber-attacks on actuators. Hence, it simultaneously regulates load voltage and distributes load current equally amongst DC-DC converters in the presence of FDI cyber-attacks.
- Unlike the finite-time control technique in [11], the proposed control approach in this paper can be applied to a general case of $N \geq 2$ parallel DC-DC converters where the converters' inductances can have different values. Furthermore, the proposed controller does not require the rate of change of output voltage that is required in [9].
- Unlike the proposed attack detection and mitigation methods (e.g., [18], [19], [21]–[23]), the proposed cooperative distributed control method does not entail significant computational burdens, as the design of resilient cooperative controllers can be done in a decentralized manner. Furthermore, we do not make any restrictive assumptions on the number of compromised DC-DC converters. As a result, the proposed resilient cooperative distributed control strategy can guarantee full resilience even if all power converters are subject to FDI cyber-attacks on their control input channels.
- We propose stability certificates based on results from network control theory and Lyapunov methods. These certificates provide theoretical guarantees of voltage regulation and balanced current sharing in the parallel interconnection of DC-DC converters, regardless of the existence of FDI attacks on actuators and the number of attacked converters. The extreme attack scenario that would make the proposed control framework invalid is also analyzed in this paper. The simulation and experimental results demonstrate the effectiveness of the proposed resilient cooperative control scheme for parallel DC-DC converters.

The rest of the paper is organized as follows: Section II formulates the problem addressed in this paper. Section III proposes a resilient cooperative distributed control mechanism that guarantees the voltage regulation and balanced current sharing among constituent converters in the presence of false data injection attacks, and discusses theoretical stability analysis aspects. The FDI-attack-resilient property of the proposed

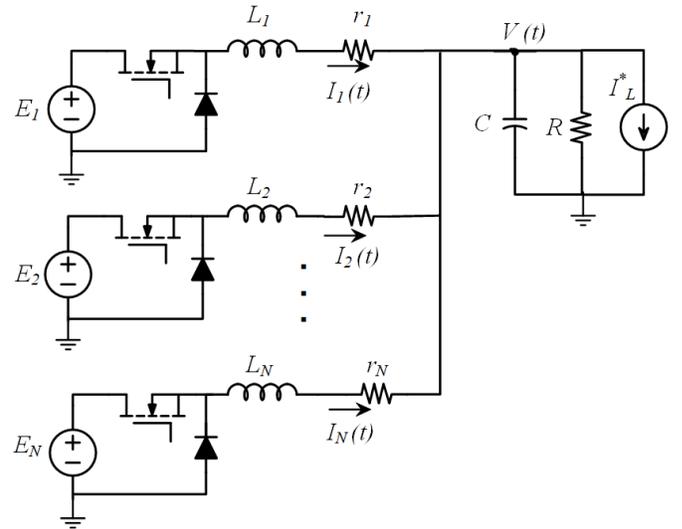


Fig. 1. The parallel connection of N DC-DC buck converters feeding a common load.

cooperative distributed control approach is investigated in Section IV and comprehensive design criteria of the control parameters of the proposed resilient distributed control framework are given. The experimental results and comparative simulation case studies are given in Section V. The paper ends with concluding remarks in Section VI.

Notation: The notation used in this paper is standard. In particular, $\mathbf{1}_N$, $\mathbf{0}_N$, and \mathbf{I}_N are an $N \times 1$ vector of ones, an $N \times 1$ zero vector, and an $N \times N$ identity matrix, respectively. For a symmetric matrix X , the positive definite and negative definite operators are shown by $X \succ 0$ and $X \prec 0$, respectively. The symbol $\text{diag}(x_1, \dots, x_N)$ indicates a diagonal matrix whose diagonal elements are x_i , $i = 1, \dots, N$.

II. PARALLEL DC-DC CONVERTERS

A. Dynamic Models

Consider a parallel interconnection of N heterogeneous DC-DC converters connected to a common load, as depicted in Fig. 1.

The dynamics of each DC-DC buck converter can be derived from Kirchhoff's current (KCL) and voltage laws (KVL), as follows:

$$\begin{aligned} L_i \dot{I}_i(t) &= -r_i I_i(t) - V(t) + u_i(t), \\ C \dot{V}(t) &= \sum_{i=1}^N I_i(t) - \frac{V(t)}{R} - I_L^*, \end{aligned} \quad (1)$$

for $i = 1, \dots, N$, where L_i is inductance of converter i , C is the output capacitance, r_i is the parasitic resistance of the inductor L_i , $R > 0$ is the common load resistance, $I_L^* \geq 0$ is a constant current load, $I_i(t)$ is the current of converter i , and $V(t)$ is the load voltage. The control input of the converter i is $u_i(t) = E_i d_i(t)$ where $d_i(t)$ is the duty cycle of converter i and E_i is the DC voltage of the input side of converter i .

B. False Data Injection Attacks on Control Channels

Malicious attackers might inject false data to perturb the local control signal $u_i(t)$. This perturbation might lead to a

detrimental impact on the stability and performance of the parallel converter system. Hence, it is required to develop a control mechanism which is resilient to such attacks.

Under the potential FDI attacks on the control input $u_i(t)$, one obtains that:

$$\hat{u}_i(t) = u_i(t) + \Delta u_i(t), \quad (2)$$

where $\hat{u}_i(t)$ is the disrupted control input applied to the converter i , $u_i(t)$ is the desired control input, and $\Delta u_i(t)$ indicates the false data injection to the control input of the converter i . Note that by injecting false data $\Delta u_i(t)$ in (2) to control input channels $u_i(t)$, incorrect control decisions are sent to each DC-DC converter, this might lead to potential equipment damages and adversely impact the performance of parallel converter systems.

Assumption 1. *It is assumed that $\Delta u_i(t)$ is a uniformly bounded signal for each converter. Moreover, the false data injection $\Delta u_i(t)$ is independent of the current and voltage signals of converters.*

Note that the assumption on uniformly bounded false data attack injections is reasonable, as unbounded FDI cyber-attacks can easily be detected by conventional anomaly detection protocols [25]. Moreover, in the case of unbounded attack injections, simple filtering can be applied so that excessively large signals received by actuators are removed or filtered [25].

C. Control Problem Statement

The main aim of this paper is to develop a control mechanism for parallel-connected DC-DC converters so that the following objectives are guaranteed:

Voltage Regulation. The first control objective is to regulate the steady-state value of DC bus voltage $V(t)$ at a given reference value V^* for the unknown load profile; i.e.,

$$\lim_{t \rightarrow \infty} V(t) = V^*. \quad (3)$$

Balanced Current Sharing. The second objective is to equally distribute the total current demand among the converters at the steady state; i.e.,

$$\lim_{t \rightarrow \infty} I_1(t) = \dots = \lim_{t \rightarrow \infty} I_N(t), \quad (4)$$

Resilience to FDI cyber-attacks on Actuators. The third objective is the resilience of the proposed control strategy against bounded FDI attacks $\Delta u_i(t)$ on the control input channels $u_i(t)$. As a result, the voltage regulation and balanced current sharing should be guaranteed regardless of the existence of such attacks.

III. PROPOSED RESILIENT CONTROL STRATEGY

This section presents a load-independent cooperative distributed control mechanism for the voltage regulation and balanced current sharing problems in the parallel-connected DC-DC converters. We show that our proposed control strategy is resilient to the bounded perturbation in the control input channels and is robust to uncertainties in the physical parameters of DC-DC converters as well as the common load.

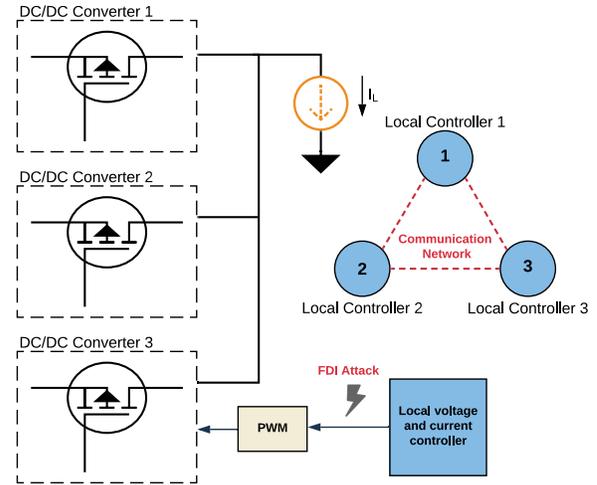


Fig. 2. A schematic diagram of a $N=3$ parallel DC-DC converter system augmented with the proposed resilient cooperative control strategy in (5). It is assumed that the control input channels are subject to false data injection attacks. In this figure, the solid black lines show the physical interconnection of DC-DC converters whereas the red dashed lines show the communication links amongst local controllers.

A. Resilient Cooperative Distributed Control Approach

In order to guarantee both voltage regulation and current sharing problems in (3) and (4), the following local voltage and current controller is proposed for converter i ; $i = 1, \dots, N$:

$$\begin{aligned} u_i(t) &= k_{i,1}V(t) + k_{i,2}I_i(t) + k_{i,3}v_i(t) + k_{i,4} \sum_{j=1}^N \alpha_{i,j} (I_i(t) - I_j(t)), \\ \dot{v}_i(t) &= -V(t) + V^* - \gamma \sum_{j=1}^N \alpha_{i,j} (I_i(t) - I_j(t)), \end{aligned} \quad (5)$$

where $\alpha_{ij} = \alpha_{ji} \geq 0$, $\gamma > 0$, and $K_i = [k_{i,1} \ k_{i,2} \ k_{i,3} \ k_{i,4}]$ are the control parameters which need to be properly designed. In (5), $v_i(t)$ is the state of the controller of converter i .

The parameters α_{ij} , $i, j = 1, \dots, N$ in (5) determine the communication amongst different converters. As one can observe from (5), the controller of each converter does not require any knowledge about the physical parameters of other converters, the capacitor value, load current, and the common load value. We will show that by a proper design of the control gains $K_i = [k_{i,1} \ k_{i,2} \ k_{i,3} \ k_{i,4}]$, the proposed cooperative control approach in (5) guarantees the closed-loop stability and provides resilience to the FDI attacks on actuators. Without these gains, the consensus-based controllers are fragile to the perturbations and false data injections to control input channels. More details about the vulnerability of existing consensus-based controllers to FDI cyber-attacks will be presented in study cases in Section V.

Remark 1. (Graph Representation of Communication Networks). *The communication network in the control strategy proposed by (5) can be represented by a connected undirected graph $\mathcal{G}_{\mathcal{E}} = (\mathcal{V}_{\mathcal{E}}, \mathcal{E}_{\mathcal{E}})$, where $\mathcal{V}_{\mathcal{E}}$ and $\mathcal{E}_{\mathcal{E}}$ are the set of vertices and edges, respectively. Each element in the vertex set $\mathcal{V}_{\mathcal{E}} = \{1, \dots, N\}$ and the edge set $\mathcal{E}_{\mathcal{E}}$ respectively represents a*

DC-DC converter and the information flow amongst existing converters. Parameters $\alpha_{i,j}$ in (5) are adjacency elements associated with the edges of the communication graph $\mathcal{G}_{\mathcal{E}}$ [26].

The overall parallel-connected buck converters in (1) with the proposed resilient cooperative control strategy in (5) can be described in a vector form as follows:

$$\begin{aligned} C\dot{\mathbf{V}}(t) &= \mathbf{1}_N^T \mathbf{I}(t) - I_L(t), \\ [L]\dot{\mathbf{I}}(t) &= ([k_1] - \mathbf{I}_N)\mathbf{1}_N V(t) + ([k_2] - [r])\mathbf{I}(t) + [k_3]\mathbf{v}(t) \\ &\quad + [k_4]\mathbb{L}_{\mathcal{E}}\mathbf{I}(t) + \Delta\mathbf{u}(t), \\ \dot{\mathbf{v}}(t) &= -\mathbf{1}_N V(t) + \mathbf{1}_N V^* - \gamma\mathbb{L}_{\mathcal{E}}\mathbf{I}(t), \end{aligned} \quad (6)$$

where $\mathbf{I}(t) = [I_1(t), \dots, I_N(t)]^T$, $\mathbf{v}(t) = [v_1(t), \dots, v_N(t)]^T$, $\Delta\mathbf{u}(t) = [\Delta u_1(t), \dots, \Delta u_N(t)]^T$, $[L] = \text{diag}(L_1, \dots, L_N)$, $[r] = \text{diag}(r_1, \dots, r_N)$, and $[k_j] = \text{diag}(k_{j,1}, \dots, k_{j,N})$ for $j = 1, \dots, 4$. Matrix $\mathbb{L}_{\mathcal{E}} \in \mathbb{R}^{N \times N}$ in (6) is the weighted Laplacian matrix associated with the undirected connected communication graph $\mathcal{G}_{\mathcal{E}}$ with an incidence matrix $\mathbb{A}_{\mathcal{E}} \in \mathbb{R}^{N \times N}$ whose (i, j) entry is $\alpha_{i,j}$ [26]. The graphical scheme of the parallel-connected DC-DC converter combined with the proposed resilient control strategy is depicted in Fig. 2.

Remark 2. (Topology of Communication Graph). The topology of the communication graph $\mathcal{G}_{\mathcal{E}}$ is free as long as it is connected and undirected. In fact, the graph $\mathcal{G}_{\mathcal{E}}$ is assumed to belong to the following set:

$$\Gamma_N = \left\{ \mathcal{G}_{\mathcal{E}} : \text{rank}(\mathbb{L}_{\mathcal{E}}) = N - 1, \mathbf{1}_N^T \mathbb{L}_{\mathcal{E}} = \mathbf{0}_N^T, \mathbb{L}_{\mathcal{E}} \mathbf{1}_N = \mathbf{0}_N \right\}. \quad (7)$$

In this paper, the following connected undirected communication network topology is used that characterizes a trade-off between performance and the number of communication links.

$$\alpha_{i,j} = \begin{cases} 1, & \text{if } |i-j|=1 \text{ or } |i-j|=N-1, \\ 0, & \text{otherwise.} \end{cases} \quad (8)$$

B. Stability Analysis

In this subsection, the stability of the closed-loop system in (6) in the presence of FDI cyber-attacks on the control signals is analyzed. To this end, the equilibria of (6) in the absence of the attack vector $\Delta\mathbf{u}(t)$ are characterized in Lemma 1 and then the stability results are presented in Lemma 2.

Lemma 1. Consider the parallel DC-DC buck converter in (1) augmented with the proposed resilient cooperative control scheme in (5) in the absence of the attack vector $\Delta\mathbf{u}(t)$. It is assumed that $k_{i,3} \neq 0$ for all for $i \in \mathcal{V}_{\mathcal{G}}$. There exists a unique equilibrium $(\bar{I}, \bar{V}, \bar{v})$ satisfying

$$\begin{aligned} \bar{I} &= \frac{1}{N} \mathbf{1}_N \left(\frac{V^*}{R} + I_L^* \right), \\ \bar{V} &= V^*, \\ \bar{v} &= [k_3]^{-1} ((\mathbf{I}_N - [k_1])\mathbf{1}_N V^* + ([r] - [k_2])\bar{I}). \end{aligned} \quad (9)$$

where \bar{V} , \bar{I} , and \bar{v} are the steady state values of the load voltage $\mathbf{V}(t)$, the converter current $\mathbf{I}(t)$, and $\mathbf{v}(t)$, respectively.

Proof. See Appendix A in Section VII. \square

Let define a new closed-loop state variable $x_{cl} = [V - \bar{V} \quad (\mathbf{I} - \bar{I})^T \quad (\mathbf{v} - \bar{v})^T]^T$. The closed-loop system in (6) can be rewritten in a state-space framework as follows:

$$\dot{x}_{cl}(t) = A_{cl}x_{cl}(t) + B_{cl}\Delta\mathbf{u}(t), \quad (10)$$

where (A_{cl}, B_{cl}) are defined as follows.

$$\begin{aligned} A_{cl} &= \begin{bmatrix} -\frac{1}{RC} & \frac{1}{C}\mathbf{1}_N^T & \mathbf{0}_{1 \times N} \\ [L]^{-1}([k_1] - \mathbf{I}_N)\mathbf{1}_N & [L]^{-1}([k_4]\mathbb{L}_{\mathcal{E}} + [k_2] - [r]) & [L]^{-1}[k_3] \\ -\mathbf{1}_N & -\gamma\mathbb{L}_{\mathcal{E}} & \mathbf{0}_{N \times N} \end{bmatrix} \\ B_{cl} &= \begin{bmatrix} \mathbf{0}_{N \times 1} & [L]^{-1} & \mathbf{0}_{N \times N} \end{bmatrix}^T. \end{aligned} \quad (11)$$

The following lemma analyzes the stability of the parallel-converter system combined with the proposed resilient control strategy in (5).

Lemma 2. For $i \in \mathcal{V}_{\mathcal{E}}$, let control gain $K_i = [k_{i,1} \quad k_{i,2} \quad k_{i,3} \quad k_{i,4}]$ belong to the following set

$$\chi_{[i]} = \left\{ (k_{i,1}, k_{i,2}, k_{i,3}, k_{i,4}) : \begin{cases} k_{i,1} < 1, & k_{i,2} < r_i \\ 0 < k_{i,3} < \frac{1}{L_i}(r_i - k_{i,2})(1 - k_{i,1}) \\ k_{i,4} = \gamma(k_{i,1} - 1) \end{cases} \right\}. \quad (12)$$

Then, the following statements hold.

- 1) A_{cl} in (11) is Hurwitz.
- 2) The states of the closed-loop system in (10) are bounded for any bounded FDI cyber-attacks $\Delta\mathbf{u}(t)$ on the control input channels.

Proof. See Appendix B in Section VII. \square

Remark 3. Note that $\chi_{[i]}$ is not an empty set since inequalities $k_{i,1} < 1$ and $k_{i,2} < r_i$ imply that $0 < \frac{1}{L_i}(r_i - k_{i,2})(1 - k_{i,1})$.

Remark 4. (Robustness to Physical System Parameters). One of the main features of the proposed control technique in (5) is its robustness against uncertainties affecting the load resistance R and capacitance C . As one can observe from (5) and (12), the control law and the closed-loop stability are independent of these parameters. Therefore, the closed-loop system in (6) is robustly stable with respect to the parameter uncertainty in R and C . If the converter inductance L_i has an interval uncertainty $L_{i,\min} \leq L_i \leq L_{i,\max}$, the control gain $k_{i,3}$ in (5) should be chosen such that $0 < k_{i,3} < \frac{1}{L_{i,\max}}(r_i - k_{i,2})(1 - k_{i,1})$. As a result, the accurate voltage regulation and current sharing are achieved for every value of $R > 0$, $C > 0$, and $L_{i,\min} \leq L_i \leq L_{i,\max}$; $i = 1, \dots, N$.

IV. ATTACK-RESILIENCE FEATURE OF PROPOSED COOPERATIVE DISTRIBUTED CONTROL STRATEGY

In this section, it is demonstrated that the proposed cooperative control approach in (5) is resilient against false data injection cyber-attacks on actuators, modeled in (2). The results are presented in the following theorem:

A. Attack-Resilience Analysis

Theorem 1. *In the presence of bounded false data injection attacks (2) on actuators, the voltage regulation and the balanced current sharing objectives can be arbitrarily accurate if $k_{i,3}, i \in \mathcal{V}_\ell$, is sufficiently large and other controller gains are selected according to Table I, so that*

$$\forall \delta_V > 0, \forall \delta_I > 0, \exists (k_{i,1}, k_{i,2}, k_{i,3}, k_{i,4}) \in \mathcal{X}_{[i]}, (i \in \mathcal{V}_\ell) : \\ \lim_{t \rightarrow +\infty} |V(t) - \bar{V}| < \delta_V, \lim_{t \rightarrow +\infty} \|\mathbf{I}(t) - \bar{\mathbf{I}}\| < \delta_I, \quad (13)$$

where δ_V and δ_I are very small positive scalars, \bar{V} , and $\bar{\mathbf{I}}$ are given in (9).

Proof. Consider the linear dynamics in (10). The closed-loop state vector $x_{cl}(t)$ can be obtained as follows:

$$x_{cl}(t) = e^{A_{cl}t} x_{cl}(0) + \int_0^t e^{A_{cl}(t-\tau)} B_{cl} \Delta \mathbf{u}(\tau) d\tau. \quad (14)$$

Therefore,

$$\begin{aligned} \lim_{t \rightarrow \infty} \|x_{cl}(t)\| &\leq \lim_{t \rightarrow \infty} \|e^{A_{cl}t} x_{cl}(0)\| + \left\| \int_0^t e^{A_{cl}(t-\tau)} B_{cl} \Delta \mathbf{u}(\tau) d\tau \right\|, \\ &\leq \lim_{t \rightarrow \infty} \left\| \int_0^t e^{A_{cl}(t-\tau)} B_{cl} \Delta \mathbf{u}(\tau) d\tau \right\|, \end{aligned} \quad (15)$$

Considering that A_{cl} is a Hurwitz matrix (see Lemma 2), $\lim_{t \rightarrow \infty} \|e^{A_{cl}t} x_{cl}(0)\| = 0$. Moreover, since $\Delta \mathbf{u}(t)$ is assumed to be uniformly bounded (see Assumption 1), there exists a constant vector $\delta_u \in \mathbb{R}^{N \times 1}$ and a positive constant τ^* such that:

$$\left\| \int_0^t e^{A_{cl}(t-\tau)} B_{cl} \Delta \mathbf{u}(\tau) d\tau \right\| \leq \left\| \int_0^t e^{A_{cl}(t-\tau)} B_{cl} \delta_u d\tau \right\|, \quad (16)$$

for all $t \geq \tau^*$, by virtue of a trivial extension of [27, Lem.2]. Taking into account (16), one can obtain that

$$\begin{aligned} \lim_{t \rightarrow \infty} \|x_{cl}(t)\| &\leq \lim_{t \rightarrow \infty} \left\| \int_0^t e^{A_{cl}(t-\tau)} B_{cl} \delta_u d\tau \right\|, \\ &\leq \left\| -A_{cl}^{-1} B_{cl} \delta_u \right\|, \end{aligned} \quad (17)$$

Due to the structure of B_{cl} in (11), $B_{cl} \delta_u = \begin{bmatrix} 0 & ([L]^{-1} \delta_u)^T & \mathbf{0}_{1 \times N} \end{bmatrix}^T$. By obtaining the inverse of A_{cl}^{-1} according to [28], it can be shown that $A_{cl}^{-1} B_{cl} \delta_u$ can be determined as follows:

$$A_{cl}^{-1} B_{cl} \delta_u = \begin{bmatrix} 0 \\ \mathbf{0}_{N \times 1} \\ [k_3]^{-1} \delta_u \end{bmatrix}. \quad (18)$$

Hence, for a large value of $k_{3,i}, \forall i \in \mathcal{V}_\ell$, $x_{cl}(t)$ converges as close to zero as desired, at the steady-state. As a result, voltage regulation and balanced current sharing objectives stated in (3) and (4) are achieved with arbitrary accuracy regardless of the existence of bounded false data injection attacks on actuators. \square

TABLE I
DESIGN CRITERIA OF COOPERATIVE DISTRIBUTED CONTROLLER PARAMETERS IN (5).

Control Parameter	Design Criteria
γ	$\gamma > 0$
$k_{i,1}$	$k_{i,1} < 1$
$k_{i,2}$	$k_{i,2} < r_i$
$k_{i,3}$	$0 < k_{i,3} < \frac{1}{L_i}(r_i - k_{i,2})(1 - k_{i,1})$ & sufficiently high
$k_{i,4}$	$k_{i,4} = \gamma(k_{i,1} - 1)$

B. Design Procedure of Proposed Resilient Cooperative Distributed Control Strategy

The design criteria for the proposed resilient cooperative distributed control strategy in (5) are summarized in Table I. According to this design guideline, the stability and resilience to FDI cyber-attacks in (2) are simultaneously ensured. As one can see from Table I, the design of control gain $K_i = [k_{i,1} \ k_{i,2} \ k_{i,3} \ k_{i,4}]$ for DC-DC converter i relies only on the local information of the converter and does not require the global knowledge about the number or parameters of other converters. Moreover, as discussed in Theorem 1, the value of $k_{i,3}$ plays an important role in the attack-resilience feature of the proposed distributed controller to FDI attacks on actuators. More specifically, the large value of $k_{i,3}$ enhances the resilience to FDI cyber-attacks on actuators.

C. An Extreme FDI Cyber-attack Scenario

Suppose that the magnitude of the false data injection attack vector $\Delta \mathbf{u}(t)$ in (10) is large. In view of (17) and (18), this might challenge the attack-resilience accuracy of the proposed distributed control strategy in (5), as δ_u in (17) will be large as well. Yet, the control inputs $u_i(t)$ are subject to a constraint $0 \leq u_i(t) \leq E_i$, so that such an attack can easily be detected by an anomaly detection algorithm [25]. As a result, the proposed FDI attack-resilient distributed control strategy can still be reliable in protecting against FDI attacks on actuators.

V. RESULTS

A. Simulation Case Studies

We consider the parallel connection of $N = 4$ DC-DC converters with different inductance values whose parameters are provided in Table II. In the simulation case studies carried out in MATLAB/Simscape environment, a switching model of DC-DC buck converters is used. The parameters of resilient distributed controller for each converter is designed based on the design criteria given in Table I.

Comparative Case Study 1: The first comparative case study evaluates the performance of the proposed resilient distributed control strategy in (5) with respect to robustness to load variations and resilience to FDI cyber-attacks on actuators modeled in (2). To this end, it is assumed that there is a load change at $t = 1.5$ s and the actuators of all four converters are subject to a mixture of constant and uniformly bounded time-varying FDI attacks launched at $t = 2$ s. The performance of the proposed resilient cooperative distributed control strategy in (5) is compared with the proposed resilient controller given

TABLE II
ELECTRICAL AND CONTROL PARAMETERS OF EXPERIMENTAL SETUP
IN FIG. 5 AND SIMULATION CASE STUDIES.

Experimental Tests	
<i>Electrical Parameters</i>	
Voltage reference V^*	48 V
Input voltage of converters E_i , $i = 1, 2, 3$	100 V
Switching frequency f_s	10 kHz
Inductance L_i , $i = 1, 2, 3$	860 μ H
Capacitance C	1100 μ F
<i>Control Parameters</i>	
γ	0.25
$K_i = [k_{i,1} \ k_{i,2} \ k_{i,3} \ k_{i,4}]$, $i = 1, 2, 3$	[-1 -1 150 -0.5]
Simulation Case Studies	
<i>Electrical Parameters</i>	
Voltage reference V^*	48 V
Input voltage of converters E_i , $i = 1, \dots, 4$	110 V
Switching frequency f_s	30 kHz
Inductance [$L_1 \ L_2 \ L_3 \ L_4$]	[1 1.5 2 1] mH
Capacitance C	1100 μ F
Load resistance R	2 Ω
<i>Control Parameters</i>	
γ	10
$K_i = [k_{i,1} \ k_{i,2} \ k_{i,3} \ k_{i,4}]$, $i = 1, \dots, 4$	[-2.5 -10 500 -35]

in [24], which is based on a smooth adaptive distributed secondary control framework. The voltage and current trajectories of the converters are depicted in Fig. 3. As one can observe from this figure, both controllers are able to mitigate the adverse impact of FDI cyber-attacks on voltage regulation and current-sharing performance. However, the proposed resilient distributed control strategy in (5) provides better and more smooth transient responses and mitigates the cyber-attacks faster than the resilient distributed controller in [24].

Comparative Case Study 2: The second comparative case study tests the performance of the proposed resilient distributed control strategy in (5) and the conventional cooperative distributed secondary control in [29] in terms of voltage regulation and current sharing performance as well as robustness to load changes. Both controllers rely on the same communication network with a connected undirected communication graph belonging to (7). The load voltage is initially regulated at 48 V and the load current is equally shared amongst all four DC-DC converters. Then, the voltage reference V^* is stepped up to $V^* = 60$ V at $t = 2$ s. Furthermore, the common load is suddenly increased at $t = 2.5$ s. The performance of both controllers is compared and the results are shown in Fig. 4. The results in Fig. 4 indicate that both controllers are able to track the reference voltage with a zero steady-state error (Fig. 4 (a), (c)); moreover, the load current is equally shared amongst participating converters (Fig. 4 (b),(d)). However, the proposed distributed control framework in (5) provides a faster and more smooth response compared to the conventional distributed control in [29] for voltage regulation and current sharing.

B. Experimental Results

Setup Description: The performance of the proposed resilient cooperative control strategy in (5) is evaluated for a case study of $N = 3$ parallel DC-DC buck converters.

The parallel converter system with the proposed resilient cooperative mechanism with the communication graph in (8) is implemented by an experimental setup, as illustrated in Fig. 5. The electrical and control parameters of the system under study are given in Table II. The proposed resilient controller is implemented in MATLAB/Simulink and run on a dSpace MicroLabBox embedded controller. To illustrate the performance of the proposed resilient control framework in (5) in terms of robust stability, resilience to FDI attacks, and voltage reference tracking, several case studies are presented.

Voltage Regulation: In order to assess the performance of the proposed control approach in terms of voltage reference tracking, it is assumed that the reference voltage V^* of the common load is initially set at 48 V. Then, it is respectively stepped down and up to 24 V and 72 V at different time instances. The voltage reference, voltage of the common load, current, and the duty cycle of the converters are shown in Fig. 6. As one can observe from this figure, the controllers provide an offset-free voltage tracking performance (Fig. 6 (a)); moreover, the load current is equally distributed amongst participating converters (Fig. 6 (b)).

Robustness to Load Uncertainty: This case study illustrates the robustness and load-independent feature of the proposed control strategy in (5) with respect to load variation and uncertainty in parasitic resistance of inductors. To this end, it is assumed that all DC-DC converters initially share a common load $R_0 = 2.13 \Omega$ and the load voltage is regulated at 48 V. Then, the load resistance value is increased and decreased at several time instances, as shown in Fig. 7 (b). The dynamic responses of DC-DC converters are depicted in Fig. 7. As illustrated in Section III, the balanced current sharing and voltage regulation are achieved regardless of the load condition.

Resilience to False Data Injection Attacks to Actuators: The final test evaluates the resilience of the proposed cooperative distributed control mechanism in (5) against false data injection attacks to control input channels. For this purpose, it is assumed that all DC-DC converters are subject to constant and time-varying FDI attacks $\Delta u_1(t) = 10$, $\Delta u_2(t) = 20 \times |\sin(\frac{2\pi}{5}(t-2))|$, launched at $t = 2$ s, and $\Delta u_3(t) = 20 \times |\sin(\frac{2\pi}{5}(t-3))|$, launched at $t = 3$ s. In order to assess the resilience of the proposed controller to other types of cyber-attacks in (2), the constant false data $\Delta u_1(t) = 10$, $\Delta u_2(t) = 20$, and $\Delta u_3(t) = 15$ are simultaneously injected to the actuators of DC-DC converters at $t = 7$ s. Hence, attackers manipulate the control commands by adding incorrect signals of a maximum $\approx 42\%$ of the nominal steady-state control commands. The voltage and current trajectories as well as the control input signals are depicted in Fig. 8 (a)-(c).

In order to highlight the superiority of the proposed resilient cooperative control in (5) in terms of resilience to cyber-attacks on actuators to non-resilient cooperative control approaches, this case study is repeated where the proposed control approach method in [15] is employed. The proposed

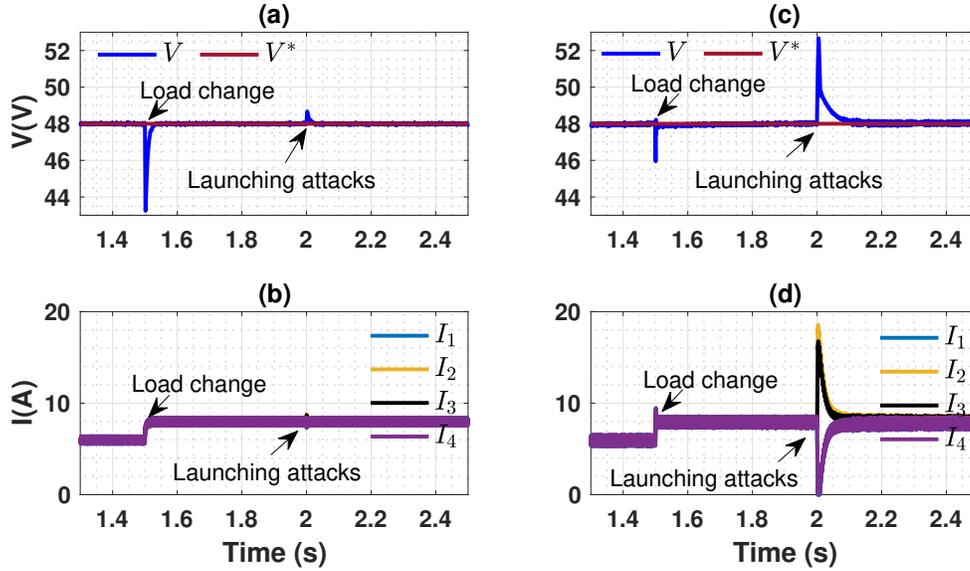


Fig. 3. Simulation results: Performance of the proposed resilient cooperative distributed controller in (5) and the resilient distributed control approach in [24] to a load change at $t = 1.5$ s and FDI attacks $\Delta u(t)$ launched at $t = 2$ s: (a),(b) voltage and current trajectories via (5) and (c),(d) voltage and current trajectories via the resilient distributed control technique in [24].

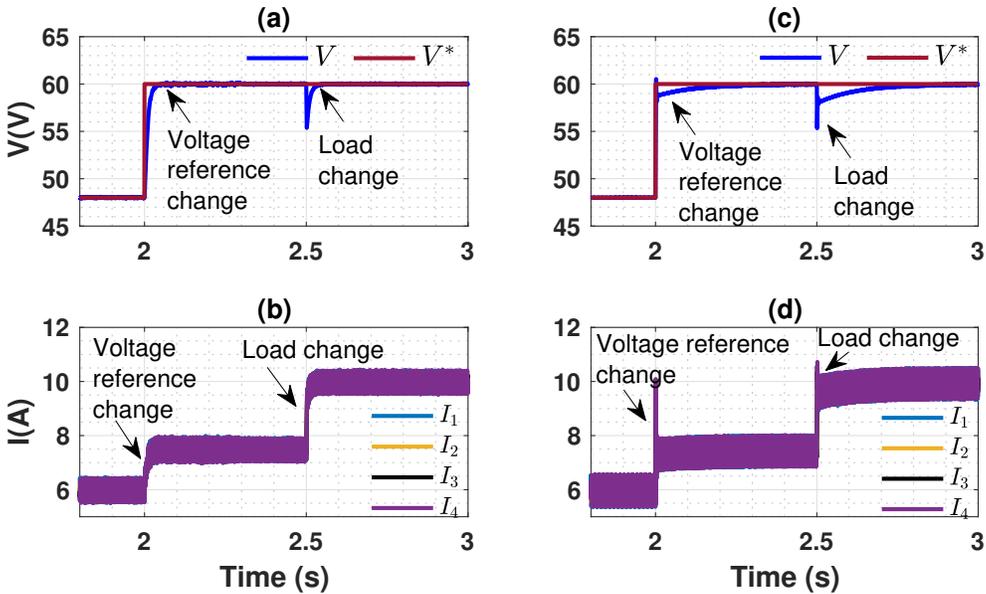


Fig. 4. Simulation results: Performance of the proposed resilient cooperative distributed controller in (5) and the conventional cooperative distributed secondary control approach in [29] to voltage reference change at $t = 2$ s and a load change at $t = 2.5$ s: (a),(b) voltage and current trajectories via (5) and (c),(d) voltage and current trajectories via the conventional distributed control technique in [29].

distributed controller in [15] for converter i is as follows:

$$\begin{aligned}
 T_{\theta_i} \dot{\theta}_i(t) &= -\sum_{j=1}^N \alpha_{ij} (I_i(t) - I_j(t)), \\
 T_{\phi_i} \dot{\phi}_i(t) &= -\phi_i(t) + I_i(t), \\
 u_i(t) &= -K_i (I_i(t) - \phi_i(t)) + \sum_{j=1}^N \alpha_{ij} (\theta_j(t) - \theta_i(t)) + V^*,
 \end{aligned} \tag{19}$$

where $\alpha_{ij} = \alpha_{ji} \geq 0$, $K_i > 0$, $T_{\theta_i} > 0$, and $T_{\phi_i} > 0$. The results are shown in Fig. 8 (d)-(f). It should be noted that the

comparison results in Fig. 8 (d)-(f) are based on MATLAB simulations.

As one can observe from Fig. 8, the proposed cooperative averaging control method in [15] is fragile to FDI cyber-attacks. As a result, any perturbation $\Delta u_i(t)$ in (2) in control input signals (actuators) leads to the failure in the voltage regulation and balanced current sharing. In contrast, as discussed in Section IV, upon launching the cyber-attacks, the proposed cooperative control mechanism in (5) mitigates the negative effects of attacks on voltage regulation and balanced current sharing. Note that the voltage fluctuations in Fig. 8 (a) are less than $\pm 0.5\%$ of the voltage reference V^* .

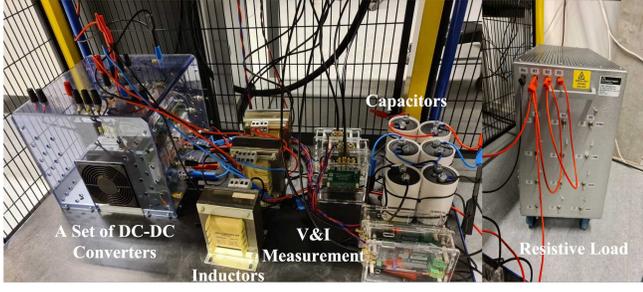


Fig. 5. Experimental setup of parallel converters comprising of three DC-DC buck converters.

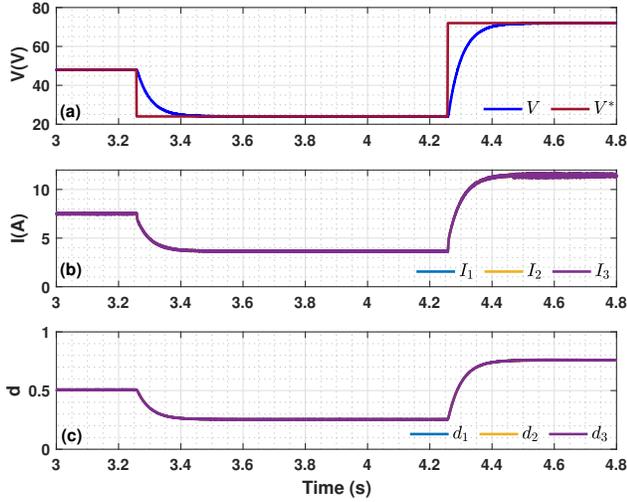


Fig. 6. Experimental results: Dynamic responses of the parallel converters in Fig. 5 to voltage reference changes: (a) common load voltage, (b) current of DC-DC converters, and (c) converters' duty cycle.

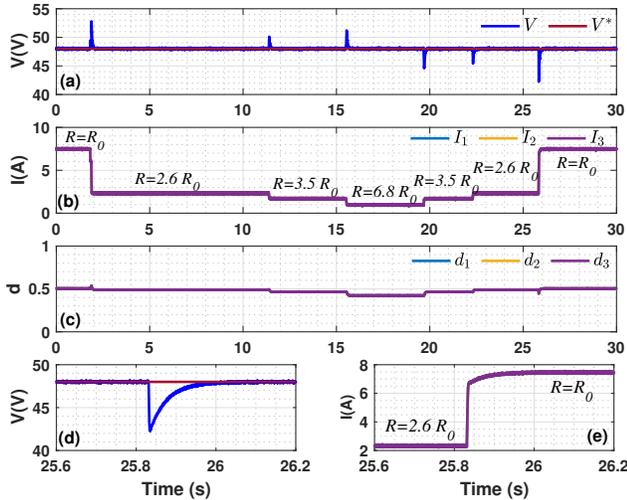


Fig. 7. Experimental results: Dynamic responses of the parallel converters to multiple load changes: (a) common load voltage, (b) current of DC-DC converters, (c) converters' duty cycle, (d) load voltage (zoomed version) for a large load change from $R = 2.6R_0$ to R_0 , and (e) converters' current signals (zoomed version) for a large load change from $R = 2.6R_0$ to R_0 - The x-axis in all subplots is time (s).

VI. CONCLUSION

Voltage regulation and balanced current sharing in the parallel connection of DC-DC converters in the presence of false data injection cyber-attacks are challenging. In this paper, we propose a resilient cooperative distributed control strategy that simultaneously regulates load voltage and distributes load current amongst active converters. The proposed cooperative distributed controller guarantees resilience against false data injection cyber-attacks in actuators and robustness with respect to uncertainties in the physical parameters of DC-DC converters as well as loads. The paper describes theoretical aspects involved in the control design, stability analysis, as well as resilience to FDI cyber-attacks and evaluates the performance of the proposed control mechanism based on comparative simulation case studies and experimental results. The future work will consider (i) the resilience to FDI attacks on communication links in cooperative and distributed control systems, (ii) the analysis of communication failures and delays in the proposed distributed control approach, and (iii) the extension of stability and attack-resilience analysis to state-dependent FDI cyber-attacks.

VII. APPENDIXES

Appendix A: Proof of Lemma 1

Consider the closed-loop system in (6). In the steady state, we have

$$0 = \mathbf{1}_N^T \bar{I} - \bar{I}_L, \quad (20a)$$

$$\mathbf{0}_N = -\mathbf{1}_N \bar{V} + \mathbf{1}_N V^* - \gamma \mathbb{L}_{\mathcal{G}} \bar{I}, \quad (20b)$$

$$\mathbf{0}_N = ([k_1] - \mathbf{I}_N) \mathbf{1}_N \bar{V} + ([k_2] - [r]) \bar{I} + [k_3] \bar{v} + [k_4] \mathbb{L}_{\mathcal{G}} \bar{I}, \quad (20c)$$

where $\bar{I}_L = \frac{\bar{V}}{R} + I_L^*$. By multiplying both sides of (20b) by $\frac{1}{N} \mathbf{1}_N^T$, one obtains that

$$\frac{1}{N} \mathbf{1}_N^T \mathbf{1}_N (-\bar{V} + V^*) - \frac{\gamma}{N} \mathbf{1}_N^T \mathbb{L}_{\mathcal{G}} \bar{I} = 0. \quad (21)$$

Since $\mathbf{1}_N^T \mathbb{L}_{\mathcal{G}} = \mathbf{0}_N^T$ [26], from the above equation one obtains that $\bar{V} = V^*$. Replacing \bar{V} with V^* in (20b) results into the following equation:

$$-\mathbf{1}_N V^* + \mathbf{1}_N V^* - \gamma \mathbb{L}_{\mathcal{G}} \bar{I} = \mathbf{0}_N. \quad (22)$$

Hence, $\mathbb{L}_{\mathcal{G}} \bar{I} = \mathbf{0}_N$ (note that $\gamma \neq 0$). Since $\mathbf{1}_N$ is an eigenvector of the Laplacian associated with a zero eigenvalue [26], $\mathbb{L}_{\mathcal{G}} \bar{I} = \mathbf{0}_N$ implies that $\bar{I} = \mathbf{1}_N i^*$, where $i^* = \frac{1}{N} \bar{I}_L$ and $\bar{I}_L = \frac{V^*}{R} + I_L^*$. From (20c), \bar{v} is obtained as follows:

$$\bar{v} = [k_3]^{-1} ((\mathbf{I}_N - [k_1]) \mathbf{1}_N V^* + ([r] - [k_2]) \bar{I}). \quad (23)$$

This completes the proof.

Appendix B: Proof of Lemma 2

Let $\Delta \mathbf{u}(t) = \mathbf{0}_N$ in (10). It suffices to show that the origin in (10) is globally asymptotically stable. To this end, we first pick any $(k_{i,1}, k_{i,2}, k_{i,3}, k_{i,4})$ in $\chi_{[i]}$ and then compute following scalars:

$$\rho_{i,1} = \frac{1}{(1 - k_{i,1} - L_i \rho_{i,2})} > 0, \quad \rho_{i,2} = \frac{k_{i,3}}{(r_i - k_{i,2})} > 0. \quad (24)$$

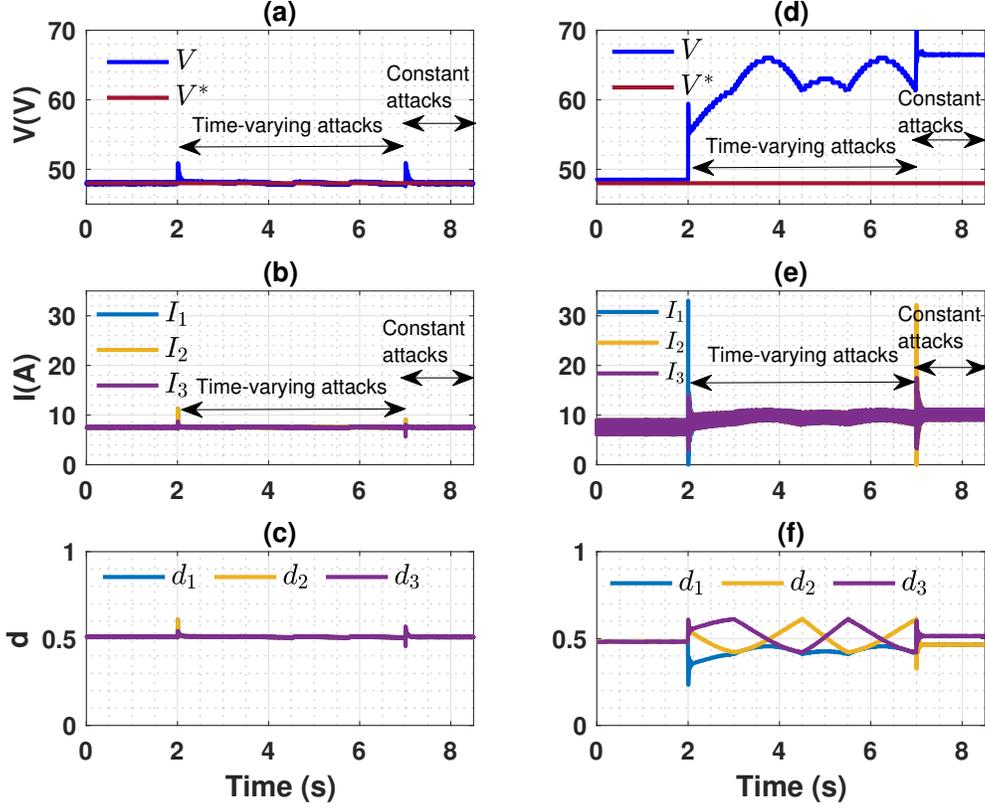


Fig. 8. Performance of the proposed resilient cooperative controller in (5) and the cooperative averaging control method in [15] to (i) time-varying FDI attacks $\Delta \mathbf{u}(t)$ launched at $t = 2$ s and $t = 3$ s and (ii) constant FDI attacks $\Delta \mathbf{u}(t)$ launched at $t = 7$ s: (a)-(c) voltage, current, and duty cycle trajectories via (5) and (d)-(f) voltage, current, and duty cycle trajectories via the non-resilient cooperative control technique in [15].

We then consider the following separable quadratic-type Lyapunov function \mathcal{V} for the parallel interconnection of N converters augmented with the resilient cooperative controllers in (5):

$$\mathcal{V} = \frac{1}{2} \sum_{i=1}^N \begin{bmatrix} e_1(t) & e_2(t) & e_3(t) \end{bmatrix} \text{diag}(C, \rho_i) \begin{bmatrix} e_1(t) \\ e_2(t) \\ e_3(t) \end{bmatrix}, \quad (25)$$

where $e_1 = V - \bar{V}$, $e_2 = \mathbf{I} - \bar{\mathbf{I}}$, $e_3 = \mathbf{v} - \bar{\mathbf{v}}$, and $\rho_i > 0$ is structured as follows:

$$\rho_i = \begin{bmatrix} L_i \rho_{i,1} & -L_i \rho_{i,2} \rho_{i,1} \\ -L_i \rho_{i,2} \rho_{i,1} & \rho_{i,2} (L_i \rho_{i,1} \rho_{i,2} + 1) \end{bmatrix}. \quad (26)$$

From this definition, it comes out that $\mathcal{V} \geq 0$. The time derivative of \mathcal{V} along the closed-loop trajectories of (6) are obtained as:

$$\begin{aligned} \dot{\mathcal{V}} &= -\frac{1}{R} e_1^2(t) + \frac{1}{2} \sum_{i=1}^N x_{cl_i}^T(t) Q_i x_{cl_i}(t) \\ &+ \frac{1}{2} \sum_{i=1}^N \left(\begin{bmatrix} e_2_i \\ e_3_i \end{bmatrix}^T \rho_i \begin{bmatrix} -1+k_{i,1} \\ L_i \\ -1 \end{bmatrix} e_4_i + e_4_i \begin{bmatrix} -1+k_{i,1} \\ L_i \\ -1 \end{bmatrix}^T \rho_i \begin{bmatrix} e_2_i \\ e_3_i \end{bmatrix} \right), \end{aligned} \quad (27)$$

where $x_{cl_i}(t)$ is defined as $x_{cl_i}(t) = [e_1(t) \ e_2_i(t) \ e_3_i(t)]^T$, $e_4_i(t) = \gamma \sum_{j=1}^N \alpha_{i,j} (e_2_i(t) - e_2_j(t))$, and

$$\begin{aligned} Q_i &= \text{diag}(1, \rho_i) \begin{bmatrix} 0 & 1 & 0 \\ \frac{-1+k_{i,1}}{L_i} & \frac{-r_i+k_{i,2}}{L_i} & \frac{k_{i,3}}{L_i} \\ -1 & 0 & 0 \end{bmatrix} \\ &+ \begin{bmatrix} 0 & 1 & 0 \\ \frac{-1+k_{i,1}}{L_i} & \frac{-r_i+k_{i,2}}{L_i} & \frac{k_{i,3}}{L_i} \\ -1 & 0 & 0 \end{bmatrix}^T \text{diag}(1, \rho_i). \end{aligned} \quad (28)$$

From (24), one obtains that $1 - k_{i,1} = \rho_{i,1}^{-1} + L_i \rho_{i,2}$. Hence, we have

$$\begin{aligned} \begin{bmatrix} e_2_i \\ e_3_i \end{bmatrix}^T \rho_i \begin{bmatrix} -1+k_{i,1} \\ L_i \\ -1 \end{bmatrix} e_4_i &= e_2_i (\rho_{i,1}(-1+k_{i,1}) + L_i \rho_{i,2} \rho_{i,1}) e_4_i \\ &+ e_3_i (-\rho_{i,1} \rho_{i,2} (-1+k_{i,1}) - \rho_{i,2} (L_i \rho_{i,1} \rho_{i,2} + 1)) e_4_i \\ &= -e_2_i e_4_i. \end{aligned}$$

Therefore, $\dot{\mathcal{V}}(e_1(t), e_2(t), e_3(t))$ can be rewritten as follows:

$$\begin{aligned} \dot{\mathcal{V}} &= -\frac{1}{R} e_1^2(t) + \frac{1}{2} \sum_{i=1}^N x_{cl_i}^T(t) Q_i x_{cl_i}(t) - \frac{1}{2} e_4^T(t) e_2(t) - \frac{1}{2} e_2^T(t) e_4(t), \\ &= -\frac{1}{R} e_1^2(t) + \frac{1}{2} \sum_{i=1}^N x_{cl_i}^T(t) Q_i x_{cl_i}(t) - \frac{\gamma}{2} e_2^T(t) (\mathbb{L}_{\mathcal{L}} + \mathbb{L}_{\mathcal{L}}^T) e_2(t), \\ &= -\frac{1}{R} e_1^2(t) + \frac{1}{2} \sum_{i=1}^N x_{cl_i}^T(t) Q_i x_{cl_i}(t) - \gamma e_2^T(t) \mathbb{L}_{\mathcal{L}} e_2(t). \end{aligned} \quad (29)$$

It is obvious that the negative semi-definiteness of \tilde{Q}_i is equivalent to the negative semi-definiteness of matrix $\tilde{Q}_i = \text{diag}(1, \rho_i^{-1}) Q_i \text{diag}(1, \rho_i^{-1})$. From the structure of ρ_i in (26), one gets:

$$\tilde{Q}_i = \begin{bmatrix} 0 & \underbrace{(L_i^{-1} \rho_{i,1}^{-1} + \rho_{i,2} - L_i^{-1} (1 - k_{i,1}))}_{\tilde{q}_{i,12}} & 0 \\ \tilde{q}_{i,12} & \frac{2}{L_i} ((-r_i + k_{i,2})(L_i^{-1} \rho_{i,1}^{-1} + \rho_{i,2}) + k_{i,3}) & \tilde{q}_{i,32} \\ 0 & \underbrace{\frac{1}{L_i} (-r_i + k_{i,2} + k_{i,3} \rho_{i,2}^{-1})}_{\tilde{q}_{i,32}} & 0 \end{bmatrix}.$$

Then, considering (24), one obtains that

$$\tilde{Q}_i = \begin{bmatrix} 0 & 0 & 0 \\ 0 & \frac{2}{L_i} (-L_i^{-1} (-r_i + k_{i,2})(1 - k_{i,1}) + k_{i,3}) & 0 \\ 0 & 0 & 0 \end{bmatrix}.$$

From (12), it comes out that $\tilde{Q}_i \preceq 0$ and, in turn, $Q_i \preceq 0$ hold. Together with $\gamma > 0$, and $\mathbb{L}_{\mathcal{L}} \succeq 0$, this proves that $\dot{\mathcal{V}} \leq 0$ holds for all $(e_1(t), e_2(t), e_3(t))$, meaning that all the state trajectories in (10) are bounded. We use the LaSalle's invariance principle to show that origin of $\dot{x}_{cl}(t) = A_{cl} x_{cl}(t)$ is globally asymptotically stable. Observe that $\dot{\mathcal{V}}(e_1^*, e_2^*, e_3^*) = 0$ is equivalent to

$$e_1^* = 0, \quad (30)$$

$$e_2^* \in \ker(\mathbb{L}_{\mathcal{L}}) \Leftrightarrow e_2^* \in \text{span}(\mathbf{1}_N), \quad (31)$$

$$x_i^* = [e_1^* \quad e_2^* \quad e_3^*]^T \in \ker(Q_i) \Leftrightarrow e_2^* - \rho_{i,2} e_3^* = 0. \quad (32)$$

Trajectories of system (10) converge to the largest invariant subset \mathcal{I} of the set defined by the above equations. From (10), $\mathbf{1}_N^T e_2^* - \frac{1}{R} e_1^* = 0$ must hold for all $(e_1^*, e_2^*, e_3^*) \in \mathcal{I}$. From (30), (31), and (32), this implies that $e_2^* = \mathbf{0}_N$ and, in turn, $e_3^* = \mathbf{0}_N$ holds in \mathcal{I} . Therefore, \mathcal{I} reduces to the origin, so that origin of $\dot{x}_{cl}(t) = A_{cl} x_{cl}(t)$ is globally asymptotically stable and A_{cl} is Hurwitz. As a result, the closed-loop system in (6) is input-to-state stable (ISS) [30]. This implies that for any bounded $\Delta u(t)$, the states of the closed-loop system are bounded too.

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